

Formulas used in ladderline loss measurements

From the Telegrapher's Equations we can derive an expression for the input impedance of a line Z_{in} , knowing the load impedance Z_l , the length of line l , the characteristic impedance of the line Z_o and the complex propagation coefficient "gamma":

$$Z_{in} = Z_o \cdot \frac{Z_l + Z_o \cdot \tanh(\gamma l)}{Z_o + Z_l \cdot \tanh(\gamma l)} \quad \text{where } \gamma = \alpha + j\beta$$

For a short-circuit termination $Z_l = 0$ so:

$$\boxed{Z_{in\text{sc}} = Z_o \cdot \tanh(\gamma l)} \quad (1)$$

For an open-circuit termination $Z_l = \infty$ so:

$$\boxed{Z_{in\text{oc}} = \frac{Z_o}{\tanh(\gamma l)}} \quad (2)$$

----- Identities & expansions -----

$$\tanh(x + y) = \frac{\tanh(x) + \tanh(y)}{1 + \tanh(x) \cdot \tanh(y)} \quad (A)$$

$$\tanh(x) = -j \tan(jx) \quad (B)$$

$$\tanh(jx) = -j \tan(-x) \quad (C)$$

$$\tanh(x) = \frac{1 - e^{-2x}}{1 + e^{-2x}} \quad (D)$$

$$\tanh(x) = x - x^3/3 + 2x^5/15 \dots \quad (E)$$

Short-circuit termination

Expanding equation (1) using identity (A) we get:

$$Z_{in\text{sc}} = Z_o \cdot \frac{\tanh(\alpha l) + \tanh(j\beta l)}{1 + \tanh(\alpha l) \cdot \tanh(j\beta l)}$$

then using identity C:

$$Z_{in\text{sc}} = Z_o \cdot \frac{\tanh(\alpha l) - j \tan(-\beta l)}{1 - j \tanh(\alpha l) \cdot \tan(-\beta l)}$$

If the line is an even number of quarter-waves long, $\tan(-\beta l) = 0$ and Z_{in} is resistive, so:

$$\boxed{R_{in} = Z_o \cdot \tanh(\alpha l)} \quad (3)$$

If the line is an odd number of quarter-waves long $\tan(-\beta l) = \pm \infty$ so for either sign:

$$\boxed{R_{in}' = \frac{Z_o}{\tanh(\alpha l)}} \quad (4)$$

where the “dash” distinguishes the high impedance case from the low impedance case.

Open-circuit termination

Expanding equation (2) with identity (A), then using (C), we get:

$$Z_{in\text{oc}} = Z_o \cdot \frac{1 - j \tanh(\alpha l) \cdot \tan(-\beta l)}{\tanh(\alpha l) - j \tan(-\beta l)}$$

If the line is an even number of quarter-waves long, $\tan(-\beta l) = 0$ and Z_{in} is resistive, so:

$$\boxed{R_{in}' = \frac{Z_o}{\tanh(\alpha l)}} \quad (5)$$

If the line is an odd number of quarter-waves long $\tan(-\beta l) = \pm \infty$ so for either sign:

$$\boxed{R_{in} = Z_o \cdot \tanh(\alpha l)} \quad (6)$$

where, again, the “dash” distinguishes the high impedance case from the low impedance case.

Determination of loss from Rin

Taking the low impedance cases - equation (3) or equation (6):

$$Rin = Zo.tanh(\alpha l)$$

Then, using identity D:

$$Rin = Zo. \frac{1 - e^{(-2\alpha l)}}{1 + e^{(-2\alpha l)}}$$

Cross multiplying:

$$Rin + Rin.e^{(-2\alpha l)} = Zo - Zo.e^{(-2\alpha l)}$$

So:

$$e^{(-2\alpha l)} = \frac{Zo - Rin}{Zo + Rin}$$

Now α is in Nepers/unit length, so $e^{(-\alpha l)}$ is the voltage attenuation along the line, and therefore $e^{(-2\alpha l)}$ is the power attenuation along the line. So loss in dB is simply given by:

$$\boxed{Loss(dB) = -10.Log \frac{Zo - Rin}{Zo + Rin}} \quad (7)$$

Alternatively, using Expansion E for $tanh(\alpha)$ and ignoring all but the first term in the series, for small losses we get the approximation:

$$Rin = Zo.tanh(\alpha l) \approx Zo. \alpha l$$

So:

$$Loss(Nepers) = \alpha l \approx Rin / Zo$$

And:

$$\boxed{Loss(dB) \approx 8.69. Rin / Zo}$$

Alternatively, taking the high impedance cases - equation (4) or equation (5):

$$Rin' = \frac{Zo}{\tanh(\alpha l)}$$

Then, using identity D: $Rin' = Zo \cdot \frac{1 + e^{(-2\alpha l)}}{1 - e^{(-2\alpha l)}}$

Cross multiplying: $Rin' - Rin' \cdot e^{(-2\alpha l)} = Zo + Zo \cdot e^{(-2\alpha l)}$

So: $e^{(-2\alpha l)} = \frac{Rin' - Zo}{Zo + Rin'}$

Now α is in Nepers/unit length, so $e^{(-\alpha l)}$ is the voltage attenuation along the line, and therefore $e^{(-2\alpha l)}$ is the power attenuation along the line. So loss in dB is simply given by:

$$\boxed{Loss(dB) = -10 \cdot \text{Log} \frac{Rin' - Zo}{Zo + Rin'}} \quad (8)$$

Or, using Expansion E for $\tanh(\alpha)$ and ignoring all but the first term in the series, for small losses we get the approximation:

$$Rin' = \frac{Zo}{\tanh(\alpha l)} \approx \frac{Zo}{\alpha l}$$

So: $Loss(Nepers) \approx \frac{Zo}{Rin'}$

And: $\boxed{Loss(dB) \approx 8.69 \cdot \frac{(Zo)}{(Rin')}}}$

We can combine (7) and (8) into a more general expression that I attribute to Owen Duffy (ex-VK1OD):

$$\boxed{Loss(dB) = -10 \cdot \text{Log} \frac{|Zo - R|}{Zo + R}} \quad (9)$$

Note: typical values for Rin are within a range that can be measured relatively accurately by simple vector impedance meters, whereas typical values for Rin' are not.

Determining Characteristic Impedance Z_o from Z_{in}

Rearranging equation (1):
$$Z_o = \frac{Z_{in\ sc}}{\tanh(\gamma l)} \quad (10)$$

Rearranging equation (2):
$$Z_o = Z_{in\ oc} \cdot \tanh(\gamma l) \quad (11)$$

Multiplying (10) by (11):
$$Z_o^2 = Z_{in\ oc} \cdot Z_{in\ sc}$$

Therefore, for any line length:

$$\boxed{Z_o = \sqrt{Z_{in\ oc} \cdot Z_{in\ sc}}} \quad (12)$$

Special case of line length = $1/8$ wavelength or odd multiples

For shorted line:
$$Z_{in} = Z_o \cdot \tanh(\gamma l)$$

Using identity (A):
$$Z_{in} = Z_o \cdot \frac{\tanh(\alpha l) + \tanh(j\beta l)}{1 + \tanh(\alpha l) \cdot \tanh(j\beta l)}$$

And identity (C):
$$Z_{in} = Z_o \cdot \frac{\tanh(\alpha l) - j \tan(-\beta l)}{1 - j \tanh(\alpha l) \cdot \tan(-\beta l)}$$

For $1/8$ wavelength (and $5/8, 9/8$ etc) $\tan(-\beta l) = -1$, so:

$$Z_{in} = Z_o \cdot \frac{\tanh(\alpha l) + j}{1 + j \tanh(\alpha l)}$$

Multiply top and bottom by conjugate $(1 - j \tanh(\alpha l))$ gives :

$$Z_{in} = Z_o \cdot \frac{(\tanh(\alpha l) + j)(1 - j \tanh(\alpha l))}{1 + \tanh^2(\alpha l)}$$

If losses are small, square terms can be ignored, so:

$$Z_{in} = Z_o \cdot \frac{\tanh(\alpha l) + j + \tanh(\alpha l)}{1}$$

Then:

$$\boxed{\text{Imag}(Z_{in}) = Z_0}$$

and:

$$\boxed{\text{Real}(Z_{in}) = 2 \cdot Z_0 \cdot \tanh(\alpha l)}$$

For 3/8 wavelength (and 7/8, 11/8 etc) $\tan(-\beta l) = +1$, so:

$$Z_{in} = Z_0 \cdot \frac{\tanh(\alpha l) - j l}{1 - j \tanh(\alpha l)}$$

Multiply top and bottom by conjugate $(1 + j \tanh(\alpha l))$ gives:

$$Z_{in} = Z_0 \cdot \frac{(\tanh(\alpha l) - j l)(1 + j \tanh(\alpha l))}{1 + \tanh^2(\alpha l)}$$

If losses are small, square terms can be ignored, so:

$$Z_{in} = Z_0 \cdot \frac{\tanh(\alpha l) - j l + \tanh(\alpha l)}{1} \quad (13)$$

Then:

$$\boxed{\text{Imag}(Z_{in}) = -Z_0}$$

and:

$$\boxed{\text{Real}(Z_{in}) = 2 \cdot Z_0 \cdot \tanh(\alpha l)}$$

So for 1/8 wavelength shorted line, and all odd multiples:

$$\boxed{Z_0 = |\text{Imag}(Z_{in})|}$$

Note: for both types of termination $\text{Real}(Z_{in})$ is the same, and is equal to twice the value for a short-circuit half-wave line or an open-circuit quarter-wave line.

Following the same methodology for a 1/8 wavelength open-circuit line we again find that:

$$\boxed{\text{Real}(Z_{in}) = 2 \cdot Z_0 \cdot \tanh(\alpha l)}$$

and:

$$\boxed{Z_0 = |\text{Imag}(Z_{in})|}$$